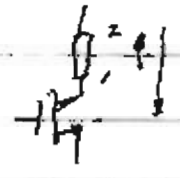
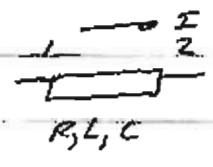


CAD #3

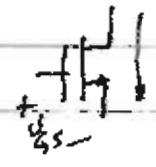
calc $2.5mA = I_D$



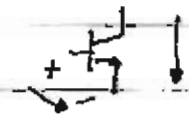
BJT as switch

MOSFET

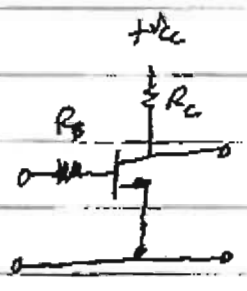
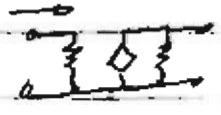
BJTS



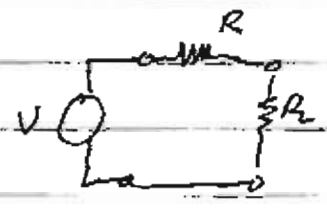
input V_{GS} output I_D
 $I = 0$



input I_B output I_C
 $V_{BE} \approx \text{constant}$
 $\approx 0.7V$



Current limiting resistor



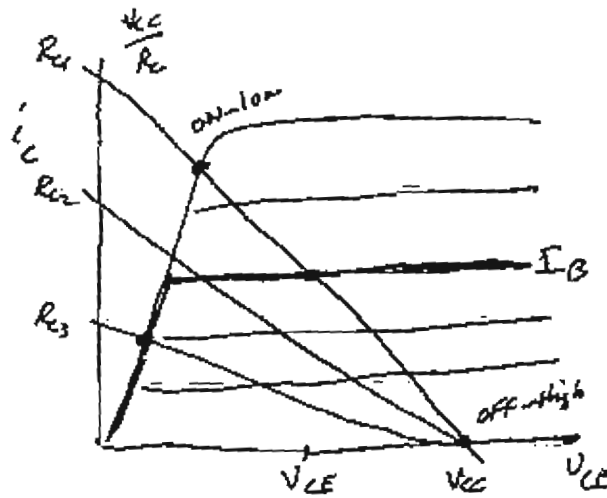
$$I_L = \frac{V}{R + R_L}$$

choose R properly

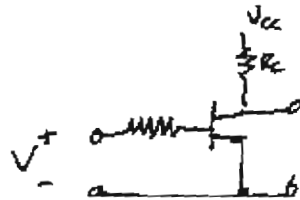
$$I_L \leq \frac{V}{R}$$

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Slope = $-\frac{1}{R_c}$



Change R_c

$$R_{c1} < R_{c2} < R_{c3}$$

$$I_B = \frac{V - V_{BE}}{R}$$

$$V_{CE} \approx 0, V_{CE} \approx 0.2$$

$$I_C = \frac{\beta(V - V_{BE})}{R}$$

$$V_{CE(sat)} = 0.2V$$

$$V_{CC} = I_C(sat) R_c + V_{CE(sat)}$$

$$R_c = \frac{V_{CC} - V_{CE(sat)}}{I_C(sat)}$$

R.O.T. $\beta_{dc(sat)} = 10$

Example: $I_C(sat) = 10 \mu A$ $V_{CC} = 5$

$\beta(sat) = 10$

$I_B(sat) = 1 \mu A$

$$R_B = \frac{V_{CC} - V_{BE}}{I_B(sat)} = \frac{5 - 0.7}{10^{-6}} = 4.3K$$

$$R_c = \frac{V_{CC} - V_{CE(sat)}}{I_C(sat)} = \frac{5 - 0.2}{10^{-2}} = 430\Omega$$

Operation

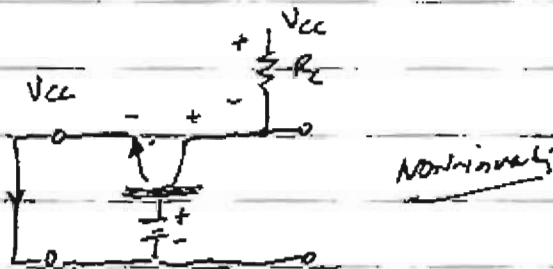
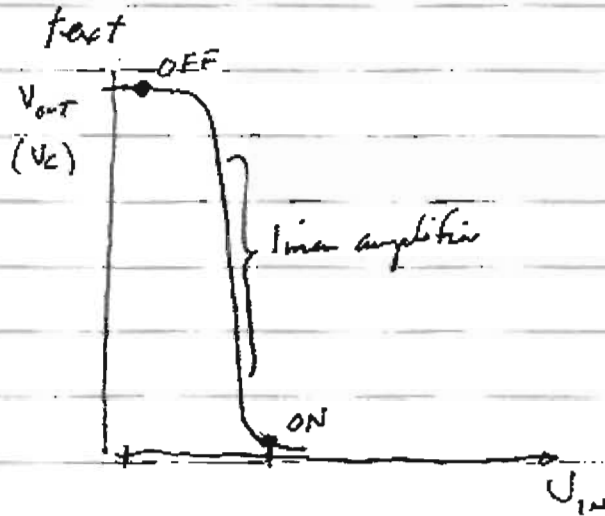
$$V_{IN} = 0,5$$

$$I_{B3} = 0,1 \text{ mA}$$

$$I_C = 0,10 \text{ mA}$$

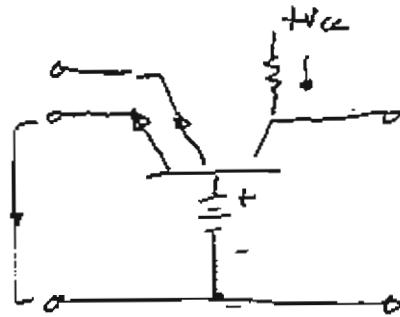
$$V_C = 5,02 \approx 0$$

Inverting Switch



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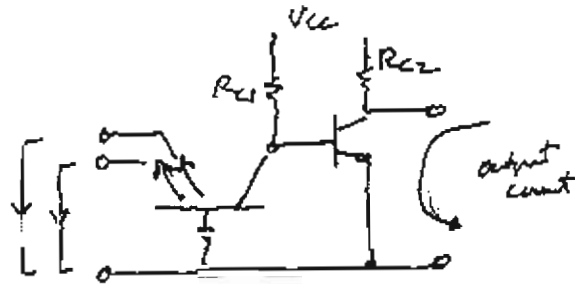


1 = high
0 = low

	input			
output	00	01	10	11
output	0	0	0	1

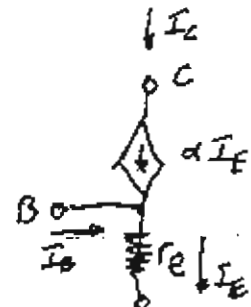
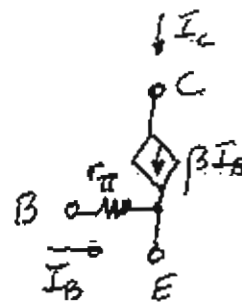
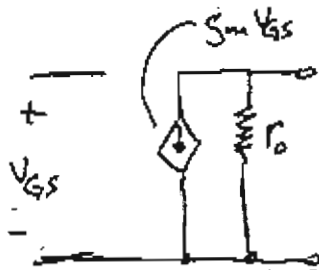
AND Gate

$$TTL = T^2 L$$

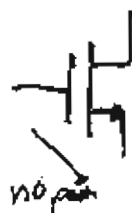


sinking current capability \equiv fan-out

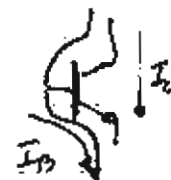
return to comparison



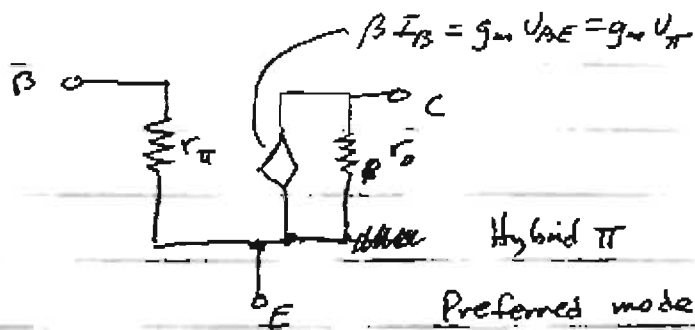
either will work T_{ec}



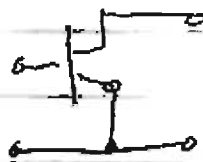
Hybrid π



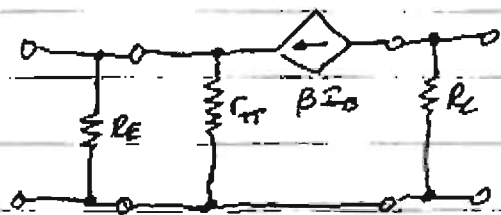
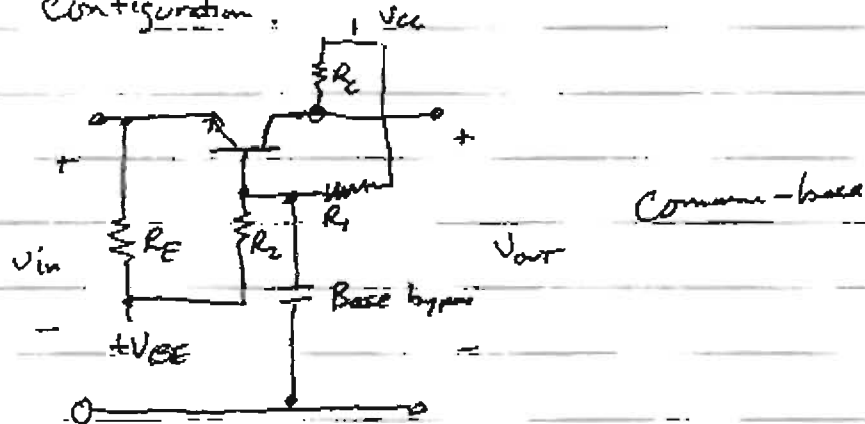
$$I_C = I_{C\alpha} = I_{E\alpha}$$



|||

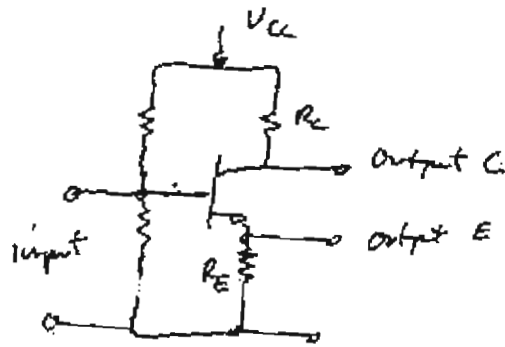


USE common-emitter model (Hybrid π) for any configuration.



Why no R_{π} ? Bypassed

Common-emitter connection using common-emitter model (Hybrid π)

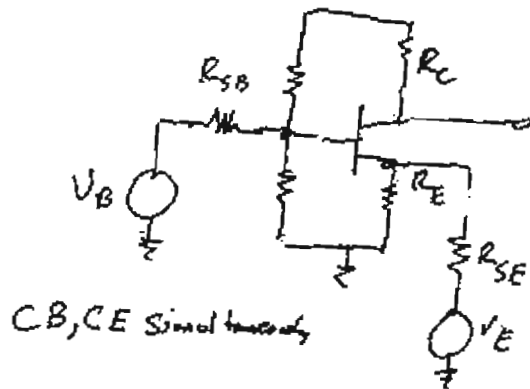


True common-emitter, bypass $R_E \Rightarrow$ no output E
only output at collector

Almost common emitter; no bypass \Rightarrow two outputs
at emitter & collector

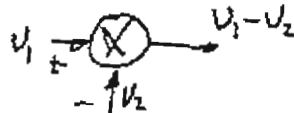
This CE, CC simultaneously.

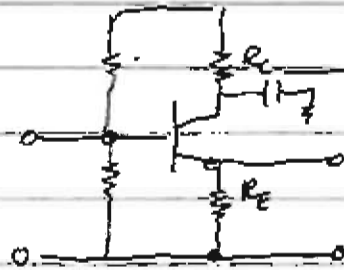
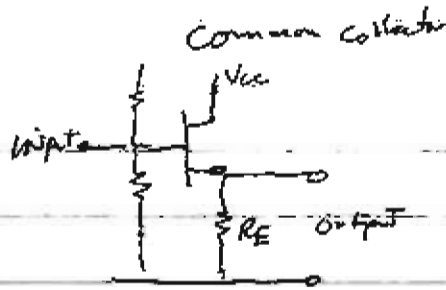
<u>INPUT</u>	<u>output(s)</u>	<u>Config</u>	<u>Uses</u>
B	E, C	CC, CE	split output
E	C	CB	current amp
C	—	—	—
E, B	C	CB, CE	difference <u>differential</u>



$$V_C = A_{V_B} V_B + A_{V_E} V_E$$

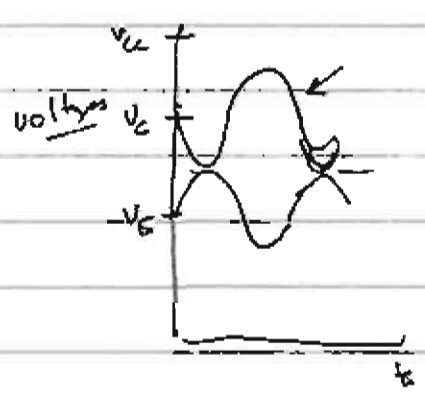
= difference between weighted inputs



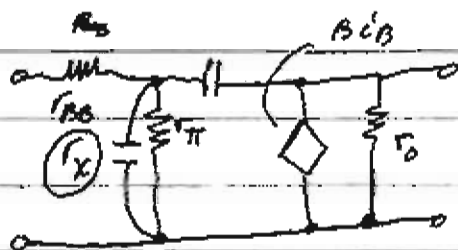


Why? dissipation in transistor
move dc power out of
transistor into RC

Suppose $V_{CC} = 15$
 $V_E = 3, V_{CE} = 6, V_{RC} = 6, I_E = 10mA$
 $P_{dc}(Q) = 60mW, P_{dc}(R_C) = 60mW$
 w/ $R_E, P_{dc}(Q) = 120mW$
 $P_d(max) = 120mW$



extension of hybrid π



LPF, STC
 affects HF response
Move To Come